

Design of a High Power X-Band Frequency Tripler Using a AlGaN/GaN HEMT Device

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Abstract— An active microwave frequency tripler using an AlGaN/GaN HEMT device is developed. This is the first reported frequency tripler implemented in GaN technology. Design of the frequency tripler is performed using a high-accuracy, multi-harmonic, wideband model which predicts the effects of self-heating and charge-trapping. A method of determining the optimal load and source networks using harmonic load-/source-pull simulations and synthesizing using harmonic reflectors is described. The tripler upconverts $f_0=3.33\text{GHz}$ to 10GHz to achieve a maximum power of $+30.0\text{dBm}$ (1.0W). As such, it provides multiplied powers which are approximately 50 times greater than those previously reported.

I. INTRODUCTION

Frequency multipliers play an important role in the generation of low noise signals by upconverting a clean signal source for use in various components of an electronic system. Systems which require multiple coherent reference signals can benefit from high efficiency frequency multiplication. While passive multiplication can be performed with great success, active frequency multipliers possess the added benefits of low conversion loss and output power, and these may potentially lead to relaxed specifications of subsequent PA stages at the upconverted frequency.

Harmonic load- and source-pull-based frequency multiplier design provides an exhaustive picture leading to the realization of an optimal network configuration and bias parameters [1]-[2]. However, the complexity of the measurement system, acquisition time and data post-processing makes such an approach an expensive option. Alternatively, CAD-based approaches are preferred [3]-[5], but may lack accuracy due to the limitations of available nonlinear active device models in predicting high-order harmonics at both input and output ports.

AlGaN/GaN (GaN) HEMT devices have become commonplace for high power, high efficiency PAs. The performance of these devices continues to improve as the fabrication technology matures. However, dispersive phenomena such as self-heating and charge-trapping are still prominent, making it difficult to characterize and produce accurate CAD models. While the manipulation of high order harmonics is a frequently employed technique in power amplifier design to improve performance [6]-[7], frequency multiplier development on GaN devices is scarce [8]-[9].

High power doublers in GaN and SiC technology have previously been demonstrated [9], but higher order frequency multipliers using wide band-gap technology have not yet been reported.

In this paper, a high-power gain frequency tripler using GaN HEMT technology is developed using a high-precision multi-harmonic active device model [10]. To the authors' knowledge, this is the first reported microwave frequency tripler using this type of device. While most of the frequency triplers found in literature produce only marginal amounts of power [11]-[20], the frequency tripler presented here produces 30.0dBm (1.0W) output power with a maximum of -2.9dB conversion gain. Section II provides an overview of the design and objectives. Section III discusses the design of the frequency tripler including the determination of bias, harmonic impedances and network realization. Section IV shows the performance of the tripler in comparison with recently published work. Finally, Section V concludes the paper.

II. OVERVIEW OF MICROWAVE FREQUENCY TRIPLER

A single-transistor, unbalanced tripler topology with the nonlinear device model representation is shown in Fig. 1. This multiplier topology is the only practical option for high power devices due to its realizability and low cost. The input and output networks (M1 and M2 in Fig. 1) will be shown to consist of reflector elements for unwanted harmonics and stub networks for input and output matching at the appropriate harmonics.

The major design objectives for the active frequency multiplier are to achieve high

- Third harmonic output power $P_{\text{out}_{3f_0}}$
- First-to-third harmonic conversion gain (CG_{3f_0})
- Suppression of unwanted fundamental f_0 ($P_{\text{out}_{1f_0}}$) and second $2f_0$ ($P_{\text{out}_{2f_0}}$) harmonics

The Cree CGH40010 10W GaN HEMT transistor [21] is used for the frequency tripler prototype. The accurate characterization and modelling of the GaN HEMT device permits the use of computerized methods to design the tripler's terminating networks. The Agilent Advanced Design System (ADS) software package is used to develop this design.

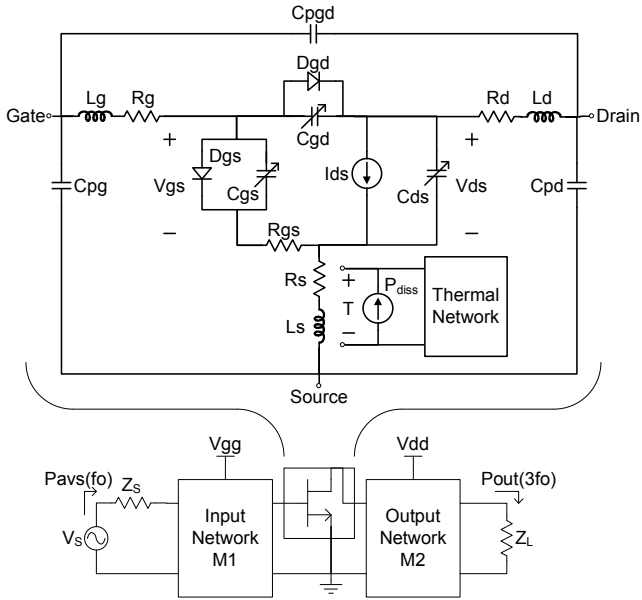


Fig. 1 Single-ended frequency tripler topology and nonlinear model.

III. FREQUENCY TRIPLER DESIGN

The design of the frequency tripler is described in this section. The key design steps consist of the device biasing, the determination of optimal harmonic terminating impedances, and the synthesis of those impedances into realizable networks. A fundamental frequency of $f_0=3.33\text{GHz}$ is used for the 10GHz GaN HEMT-based frequency tripler.

A. Device Biasing

The fundamental concept of harmonic generation is to drive the device such that its output drain current waveform becomes clipped due to the transition between operating regimes. According to Fourier analysis, a symmetrically-clipped drain current waveform is desirable in the case of a frequency tripler due to its rich odd harmonic content. The two hard nonlinearities which create this clipping effect are delineated by the pinchoff regime and the forward conduction of the gate-source diode. The ease with which the waveform is clipped and the amount of symmetry is determined primarily by the device bias, drive level and terminating impedances at each harmonic.

Double-sided clipped waveform analysis has shown that the optimum bias condition for third harmonic generation is midway between forward conduction and pinchoff [15]. Low power devices which do not suffer from significant self-heating can be conveniently biased in this manner; however, in the case of the GaN HEMT device, the IV characteristics change as a function of the dissipated power. This varies with the quiescent bias and dissipated power level as shown in the pulsed-IV (PIV) characteristics of Figs. 2. Therefore, selection of the gate bias is not a trivial choice and should be chosen utilizing extensive simulations of the output power level with respect to the gate-biasing. The bias which provides the maximum $3f_0$ output power has been determined to be $V_{gs}=-2.3\text{V}$ with a drain bias of $V_{ds}=28\text{V}$.

GaN HEMT devices are particularly well suited for third harmonic generation at the higher frequencies because they possess (1) a relatively high pinchoff voltage as compared with SiC MESFET devices and (2) a high f_T unity gain frequency. A high pinchoff voltage allows the drain current to clip with less voltage swing at the gate. A high f_T frequency allows the $3f_0$ harmonic generated at the input to be delivered to the output.

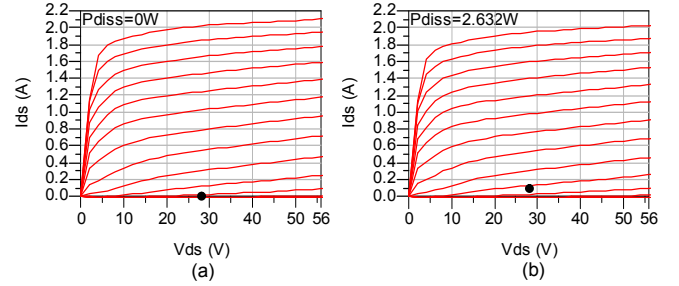


Fig. 2 Simulated PIV characteristics of the GaN HEMT model at a bias of (a) $V_{gsq}=-3.0\text{V}$, $V_{dsq}=28\text{V}$ and (b) $V_{gsq}=-2.3\text{V}$, $V_{dsq}=28\text{V}$.

B. Determination of Input and Output Harmonic Terminations

Once the application of the optimal bias level is established, proper load and source impedances are applied to achieve increased gain and output power. Harmonic load and source-pull simulations are performed on the nonlinear model to determine the optimal impedances for maximum $3f_0$ generation. Figure 3 shows the overall simulation setup. This simulation is unique in that it employs ideal triplexer components which permit the separation and independent tuning of the harmonic impedances seen by the device at both the input and output ports. This is an extension of the concept employed in [9] in that three harmonics are now considered instead of two.

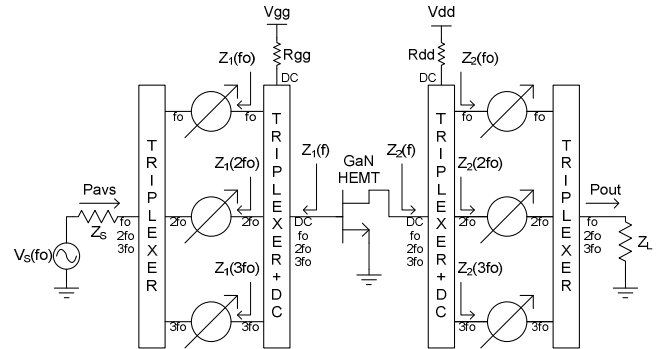


Fig. 3 Generalized harmonic load and source-pull simulation setup used for the analysing the power of the GaN HEMT device.

This generalized method is advantageous in that the effect of each harmonic load can be explored independently since the configuration assumes no particular topology. Once the harmonic impedances are determined, the designer can choose one of many possible realizations of the networks. For the source network, an impedance match at f_0 and harmonic reflectors for $2f_0$ and $3f_0$ are desirable for increasing the

overall gain and output power. Likewise, for the load network, an output match at $3fo$ provides increased power while harmonic reflectors for $1fo$ and $2fo$ provide increased gain, power and suppression of the undesirable harmonics.

The harmonic impedance values determined from this analysis that result in producing the maximum $3fo$ power are plotted in Fig. 4.

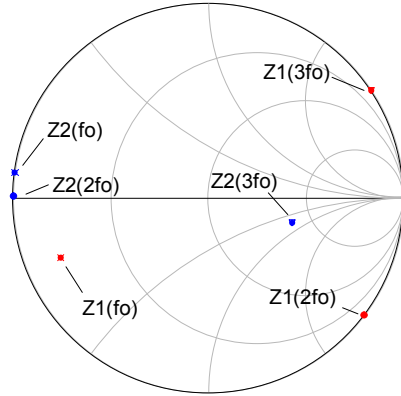


Fig. 4 Harmonic load and source impedances determined from simulation

C. Synthesis of Input and Output Networks

This prototype design uses external bias tees and, therefore, attention will be focused on the design and implementation of the microwave load and source networks. The major challenge in building a frequency tripler is realizing an output network such that a short circuited fundamental reflector does not inadvertently short circuit the desired third harmonic signal. Several authors have used combinations of transmission line (TL) and lumped elements [17],[19],[20], filters [11], coupled lines [2] and novel composite right-hand/left-hand TLs [13] to solve this problem.

One goal in the synthesis and fabrication of the harmonic networks is to minimize circuit and substrate losses. Thus the resulting topology is kept as simple as possible. The design presented here, shown in Fig. 5, utilizes a cascaded harmonic network using short feed lines at the input and output and quarter-wavelength harmonic impedance reflectors where possible. The input network consists of a matching stub for fo and reflectors for the $2fo$ and $3fo$ harmonics. The output network consists of a $2fo$ reflecting stub followed by a fundamental-rejecting coupled line section and $3fo$ matching stub. The authors have chosen to use a coupled line to pass the third harmonic, while rejecting the fundamental.

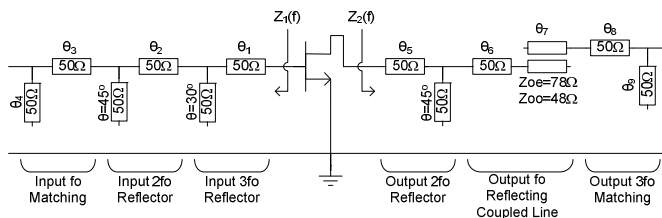


Fig. 5 Circuit topology of GaN frequency tripler prototype

The input and output networks shown in Figure 4 were fabricated in Rogers RT/Duroid 5880 substrate. The choice to use 50 ohm lines for all the transmission line elements is dictated by the necessity to avoid excessive heating of the microstrip traces resulting from high power reflection.

IV. MEASUREMENT AND PERFORMANCE

The output power of the frequency tripler was measured in a large-signal test setup capable of delivering up to +39dBm of available power.

The harmonic output power and conversion gain of the GaN HEMT frequency tripler is shown in Figs. 6 and 7, respectively. The measured output shows that the realized circuit is capable of producing 30.0dBm (1.0W) output power and 2.9dB conversion loss at the third harmonic. The simulated output closely predicts the measured $3fo$ harmonic over a reasonable range of available powers (Fig. 6). Conversion loss (Fig. 7) is also estimated closely over a useful range of available powers.

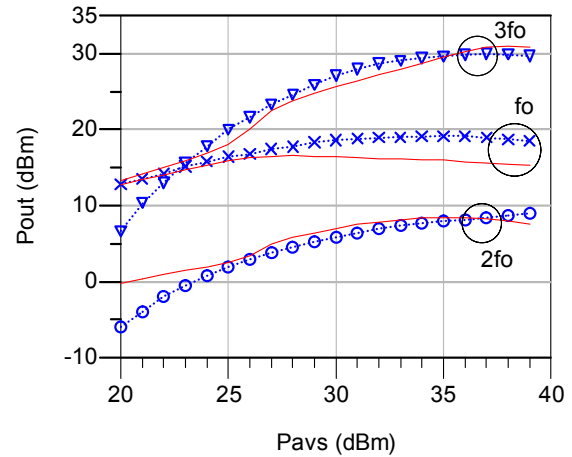


Fig. 6 Measured and simulated output power for the frequency tripler for three harmonics.

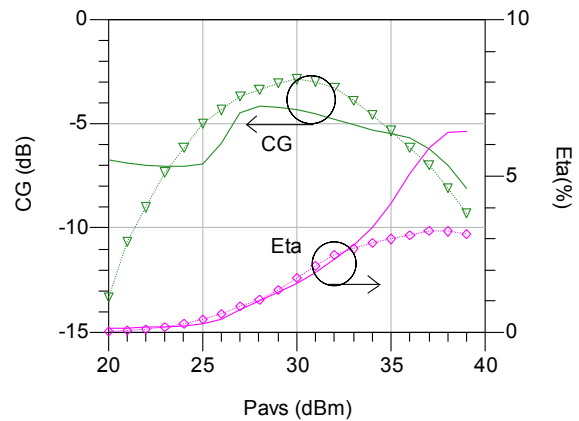


Fig. 7 Measured and simulated conversion gain and efficiency of frequency tripler

A comparison table between this work and recently published frequency triplers is shown in Table I. The table summarizes the maximum performance reported in each of the publications. According to the table, the frequency tripler design presented here represents the highest output power, single-transistor tripler available to date producing over 50 times more power than that of the next highest.

TABLE I
COMPARISON OF RECENTLY PUBLISHED TRIPLERS

Work	3fo (GHz)	CG (dB)	Pout (dBm)	η (%)	Pout _{1f₀} /Pout _{2f₀} (-dBc)
This	10	-2.9	30.0	3.14	-11.2/-21.8
[2]	7.5	-2.4	1.6	13	-33.6/-29.6
[4]	38.64	-2.7	4.7	11	-38/-23
[5]	140	-11	-1.5	—	—/—
[11]	34.5	-6.5	-6.5	—	—/—
[12]	3	5.5	6	57	<-25/<-25
[13]	3	-5.67	-5.67	—	-62.72/-53.6
[14]	9	2.9	12.9	7.5	-28/-28
[15]	42-51	—	12.5	—	-40/-20
[16]	6	0.5	6	—	-50/-19
[17]	2.475	9	12	22.5	—/—
[18]	30	-5	12	5	<-50/<-40
[19]	8.85	2.9	7.9	—	<-35/<-50
[20]	8.82	3.67	9.17	—	-28.67/-26.87

V. CONCLUSIONS

A GaN HEMT-based frequency tripler producing a maximum output of 30.0dBm and -2.9dB conversion gain has been demonstrated. The design of the frequency tripler was performed using harmonic load-/source-pull simulations utilizing a high-precision nonlinear device model. Frequency multiplication techniques using impedance matching at the desired harmonics and reflection at unwanted harmonics allow increased output power while minimizing conversion loss.

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